



## Real Time Clock with I2C or SPI, Crystal Temperature Compensation, Battery Switchover and Trickle Charger

### Description

The EM3027 is an Ultra Low Power CMOS real-time clock IC with two serial interface modes: I2C or SPI. The interface mode is selected by the chip version (see §12).

The basic clock is obtained from the 32.768 kHz crystal oscillator. A thermal compensation of the frequency is based on the temperature measurement and calculation of the correction value. The temperature can be measured internally or be input by an external application to the register.

The chip provides clock and calendar information in BCD format with alarm possibility. The actual contents are latched at the beginning of a read transmission and afterwards data are read without clock counter data corruption.

An integrated 16-bit timer can run in Zero-Stop or Auto-Reload mode.

An interrupt request signal can be provided through INT/IRQ pin generated from alarm, timer, voltage detector and Self-Recovery system.

An integrated trickle charger allows recharging backup supply  $V_{Back}$  from the main supply voltage  $V_{CC}$  through internal resistor(s). The internal device supply will switchover to  $V_{CC}$  when  $V_{CC}$  is higher than  $V_{Back}$  and vice versa.

The device operates over a wide 1.4 to 5.5V supply range and requires only 900 nA at 5V. It can detect internally two supply voltage levels.

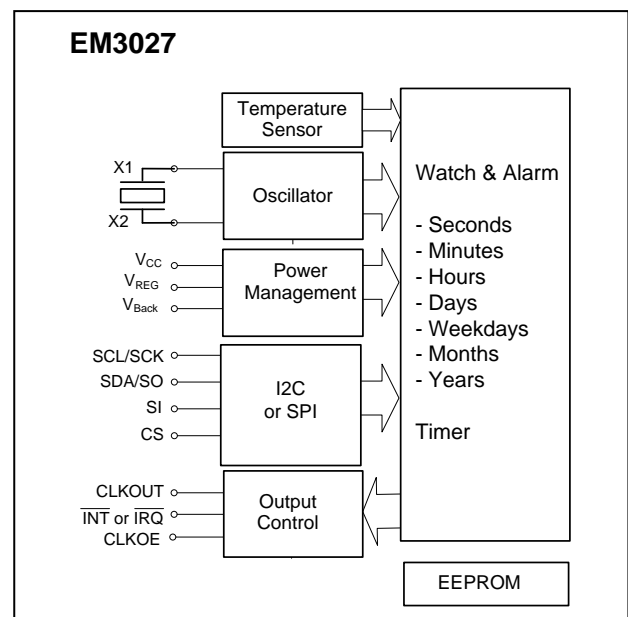
### Applications

- ☐ Utility meters
- ☐ Battery operated and portable equipment
- ☐ Consumer electronics
- ☐ White/brown goods
- ☐ Pay phones
- ☐ Cash registers
- ☐ Personal computers
- ☐ Programmable controller systems
- ☐ Data loggers

### Features

- ☐ Fully operational from 2.1 to 5.5V
- ☐ Supply current typically 600 nA at 1.4V
- ☐ Thermal compensated crystal frequency
- ☐ Oscillator stability 0.5 ppm / Volt
- ☐ Counter for seconds, minutes, hours, day of week, date months, years in BCD format and alarm
- ☐ Leap year compensation
- ☐ 16-bits timer with 2 working modes
- ☐ Two low voltage detection levels  $V_{Low1}$ ,  $V_{Low2}$
- ☐ Automatic supply switchover
- ☐ Serial communication via I2C (I<sup>2</sup>C-bus specification Rev. 03 compatible – see §10.2) or SPI (3-line SPI-bus with separate combinable data input and output)
- ☐ Thermometer readable by the host
- ☐ Trickle charger to maintain battery charge
- ☐ Integrated oscillator capacitors
- ☐ Two EEPROM and 8 RAM data bytes for application
- ☐ Digital Self-Recovery system
- ☐ No busy states and no risk of corrupted data while accessing
- ☐ One hour periodical configuration registers refresh
- ☐ Support for standard UL1642 for Lithium batteries
- ☐ Standard temperature range: -40°C to +85°C
- ☐ Extended temperature range: -40°C to +125°C
- ☐ Packages: TSSOP8, TSSOP14, SO8.

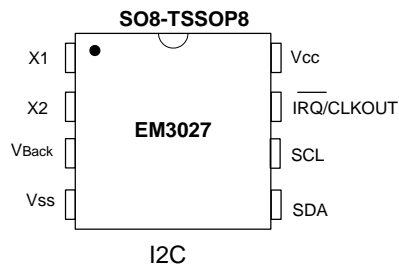
### Block Diagram



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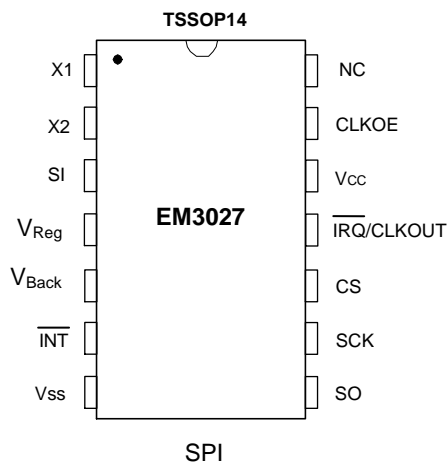
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## 1 Packages / Pin Out Configuration



Pin	Name	Function
1	X1	32.768 kHz crystal input
2	X2	32.768 kHz crystal output
3	V <sub>Back</sub>	Backup Supply
4	V <sub>ss</sub>	Ground Supply
5	SDA	Serial Data
6	SCL	Serial Clock
7	IRQ/CLKOUT	Interrupt Request/Clock output
8	V <sub>cc</sub>	Positive Supply

**Table 1**



Pin	Name	Function
1	X1	32.768 kHz crystal input
2	X2	32.768 kHz crystal output
3	SI	Serial Data input
4	V <sub>Reg</sub>	Regulated Voltage – Reserved for test purpose (This output must be left unconnected)
5	V <sub>Back</sub>	Backup Supply
6	INT	Interrupt Request output (Open Drain active low)
7	V <sub>ss</sub>	Ground Supply
8	SO	Serial Data output
9	SCK	Serial Clock input
10	CS	Chip Select input
11	IRQ/CLKOUT	Interrupt Request/Clock output
12	V <sub>cc</sub>	Positive Supply
13	CLKOE	Clock Output Enable CLKOE = '0' CLKOUT is low CLKOE = '1' CLKOUT is output
14	NC	Not Connected

**Table 2**

## 2 Absolute Maximum Ratings

Parameter	Symbol	Conditions
Maximum voltage at $V_{CC}$	$V_{CCmax}$	$V_{SS} + 6.0V$
Minimum voltage at $V_{CC}$	$V_{CCmin}$	$V_{SS} - 0.3V$
Maximum voltage at any signal pin	$V_{max}$	$V_{CC} + 0.3V$
Minimum voltage at any signal pin	$V_{min}$	$V_{SS} - 0.3V$
Maximum storage temperature	$T_{STOmax}$	+150°C
Minimum storage temperature	$T_{STOmin}$	-65°C
Electrostatic discharge maximum to MIL-STD-883C method 3015.7 with ref. to $V_{SS}$	$V_{Smax}$	2000V

**Table 3**

Stresses above these listed maximum ratings may cause permanent damages to the device.

Exposure beyond specified operating conditions may affect device reliability or cause malfunction.

### 2.1 Handling Procedures

This device has built-in protection against high static voltages or electric fields; however, anti-static precautions must be taken as for any other CMOS component. Unless otherwise specified, proper operation can only occur when all terminal voltages are kept within the voltage range. Unused inputs must always be tied to a defined logic voltage level.

## 2.2 Operating Conditions

Parameter	Symbol	Min	Typ	Max	Unit
Operating Temp.	$T_A$	-40		+125	°C
Supply voltage (Note 1)	$V_{CC}, V_{Back}$	1.4	5.0	5.5	V
Capacitor at $V_{CC}, V_{Back}$	$C_D$		100		nF

**Table 4**

Note 1: Refer to paragraphs 9.1 and 9.2

## 2.3 Crystal characteristics

Parameter	Symbol	Min	Typ	Max	Unit
Frequency	$f$		32.768		kHz
Load capacitance	$C_L$	7	8.2	12.5	pF
Series resistance	$R_S$		70	110	kΩ

**Table 5**

**Crystal Reference :** Micro Crystal CC5V-T1A  
web: [www.microcrystal.com](http://www.microcrystal.com)

## 2.4 EEPROM Characteristics

Parameter	Symbol	Min	Typ	Max	Unit
Read voltage	$V_{Read}$	1.4			V
Programming Voltage	$V_{Prog}$	2.2			V
EEPROM Programming Time	$T_{Prog}$			30	ms
Write/Erase Cycling		5000			cycles

**Table 6**

## 3 Electrical Characteristics

Parameter	Symbol	Test Conditions	$V_{CC}$	Temp. °C	Min	Typ	Max	Unit
Total supply current with Crystal	$I_{CC}$	All outputs open, $R_S < 70$ kΩ, $V_{Back} = 0V$ I2C: SDA, SCL at $V_{CC}$ , Clk/Int='0' SPI: All inputs at $V_{SS}$	1.4	-40 to 125		0.6	4.6	μA
			3.3	-40 to 125		0.8	5.2	
			5.0	-40 to 125		0.9	5.5	
Total supply current with Crystal	$I_{Back}$	All outputs open, $R_S < 70$ kΩ, $V_{CC} = 0V$ , $V_{Back} = 3.3V$ I2C: SDA, SCL at $V_{Back}$ , Clk/Int='0' SPI: All inputs at $V_{SS}$	0	-40 to 125		0.8	5.2	μA
Dynamic current I2C	$I_{DD}$	SCL = 100kHz (See Note 1)	1.4	-40 to 125			15	μA
		SCL = 400kHz (See Note 1)	3.3	-40 to 125			40	
		SCL = 400kHz (See note 1)	5.0	-40 to 125			60	

Parameter	Symbol	Test Conditions	V <sub>CC</sub>	Temp. °C	Min	Typ	Max	Unit
Dynamic current SPI Interface	I <sub>DD</sub>	SCK = 200 kHz (See Note 2)	1.4	-40 to 125			18	µA
		SCK = 1 MHz (See Note 2)	3.3	-40 to 125			55	
		SCK = 1 MHz (See Note 2)	5.0	-40 to 125			75	
Low supply detection level1	V <sub>low1</sub>	Relative to V <sub>CC</sub>		-40 to 125	1.8		2.1	V
Low supply detection level2	V <sub>low2</sub>	Relative to V <sub>CC</sub>		-40 to 125	1.0		1.4	V
Switchover hysteresis	V <sub>hyst</sub>	V <sub>CC</sub> with respect to V <sub>Back</sub> = 3.0V		-40 to 125		20		mV
Input Parameters								
Low level input voltage	V <sub>IL</sub>	CS, CLKOE, SI, SCL/SCK, SDA	1.4 to 5.0	-40 to 125			0.2V <sub>CC</sub>	V
High level input voltage	V <sub>IH</sub>			-40 to 125	0.8V <sub>CC</sub>			
Input Leakage	I <sub>IN</sub>	0.0 < V <sub>IN</sub> < V <sub>CC</sub>	1.4 to 5.0	-40 to 125	-1.5		1.5	µA
Output Parameters								
Low level output voltage	V <sub>OL</sub>	I <sub>OL</sub> = 0.4 mA	1.4	-40 to 125			0.2	V
High level output voltage	V <sub>OH</sub>	I <sub>OH</sub> = 0.1 mA			1.0			
Low level output voltage	V <sub>OL</sub>	I <sub>OL</sub> = 1.5 mA	3.3	-40 to 125			0.25	V
High level output voltage	V <sub>OH</sub>	I <sub>OH</sub> = 1.5 mA			2.7			
Low level output voltage	V <sub>OL</sub>	I <sub>OL</sub> = 5.0 mA	5.0	-40 to 125			0.8	V
High level output voltage	V <sub>OH</sub>	I <sub>OH</sub> = 2.0 mA			4.5			
Output HiZ leakage on INT	I <sub>LEAK_OUT</sub>	INT not active	1.4 to 5.0	-40 to 125	-1.5		1.5	µA
Oscillator								
Start-up voltage	V <sub>STA</sub>	T <sub>STA</sub> < 10s		-40 to 125	1.2			V
Start-up time	T <sub>STA</sub>		5.0	-40 to 125		1	3	s
Frequency stability over voltage	Δf/(f ΔV)	1.8V ≤ V <sub>CC</sub> ≤ 5.5V, T <sub>A</sub> = +25°C		25		0.5	2	ppm/ V
Input capacitance on X1	C <sub>IN</sub>	T <sub>A</sub> = +25°C, f = 32.768kHz, V <sub>meas</sub> = 0.3V (Note 3)		25		16.5		pF
Output capacitance on X2	C <sub>OUT</sub>	T <sub>A</sub> = +25°C, f = 32.768kHz, V <sub>meas</sub> = 0.3V (Note 3)		25		15.0		
Trickle Charger								
Current limiting Resistors	R80k	V <sub>CC</sub> =5.0V, V <sub>Back</sub> =3.0V		25		80		kΩ
	R20k	V <sub>CC</sub> =5.0V, V <sub>Back</sub> =3.0V		25		20		
	R5k	V <sub>CC</sub> =5.0V, V <sub>Back</sub> =3.0V		25		5.0		
	R1.5k	V <sub>CC</sub> =5.0V, V <sub>Back</sub> =3.0V		25		1.5		
Thermometer								
Precision	T <sub>E</sub>	V <sub>low1</sub> < V <sub>CC</sub> ≤ 5.5V		40		+/- 1	+/- 2	°C
				-40 to 125		+/- 4	+/- 10	

**Table 7**

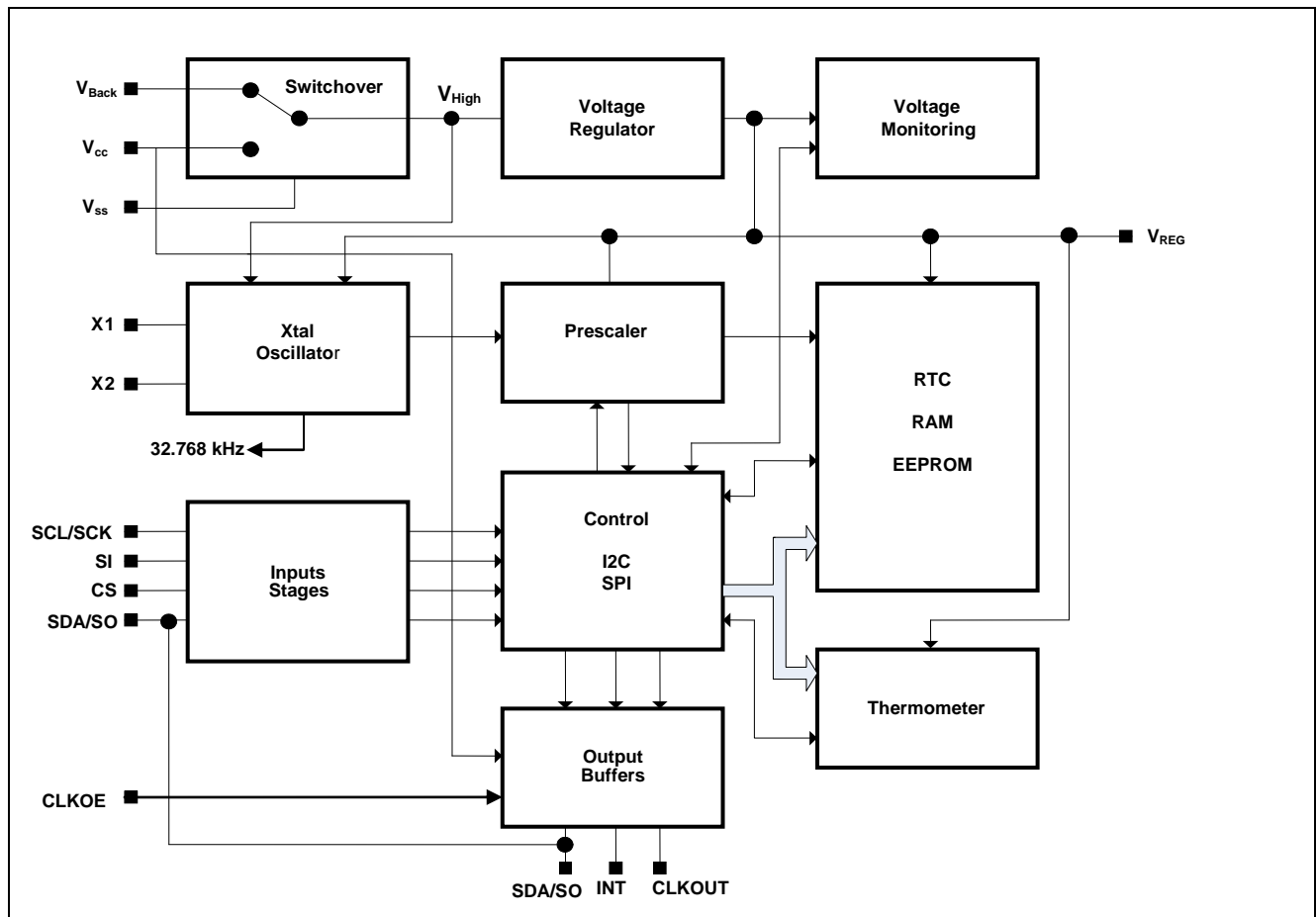
The following parameters are tested during production test: I<sub>DD</sub>, V<sub>low1</sub>, V<sub>low2</sub>, V<sub>IL</sub>, V<sub>IH</sub>, V<sub>OL</sub>, V<sub>OH</sub>, I<sub>IN</sub>, I<sub>LEAK\_OUT</sub>  
The parameters I<sub>CC</sub>, V<sub>hyst</sub>, V<sub>STA</sub>, T<sub>STA</sub>, C<sub>IN</sub>, C<sub>OUT</sub>, Δf/(f\*ΔV), T<sub>E</sub> are characterised during the qualification of the IC.

**Notes:**

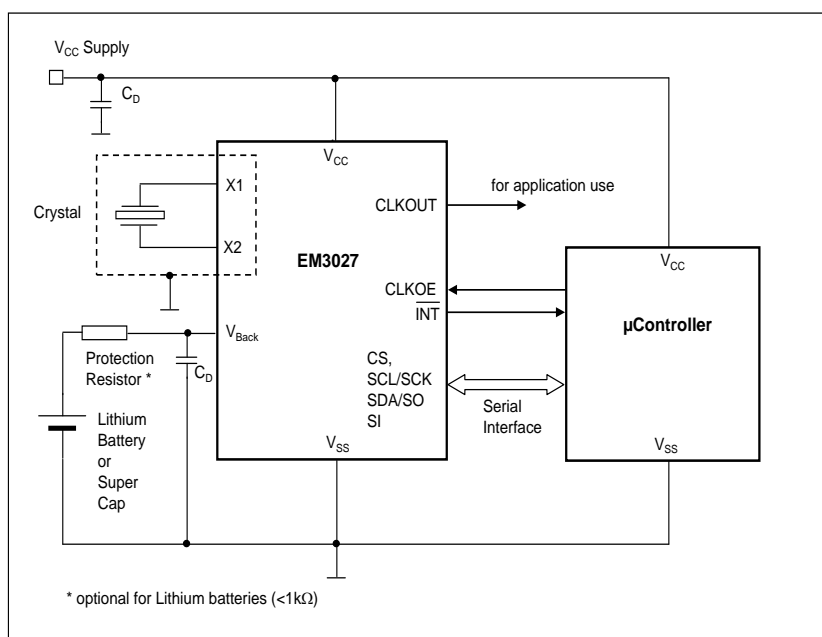
- SDA = V<sub>SS</sub>, continuous clock applied at SCL (V<sub>IL\_SCL</sub> < 0.05V, V<sub>IH\_SCL</sub> > 0.95V<sub>CC</sub>)
- CS, SI = V<sub>CC</sub>, continuous clock applied at SCK, SO not connected. (V<sub>IL\_SCK</sub> < 0.05V<sub>CC</sub>, V<sub>IH\_SCK</sub> > 0.95V<sub>CC</sub>)  
Note that there is a 100kΩ pull-down resistor on CS.
- V<sub>meas</sub>: Peak to peak amplitude during capacitance measurement
- Below 0°C, a negative slope on V<sub>CC</sub> will stop the oscillator during a time which may be as long as 1 second. This is not dependent on slope and occurs with a slew rate as low as 2V/sec.
- Some chip versions do not support "Trickle Charger" and "Switchover" features.

## 4 EM3027 Block Diagram and Application Schematic

### 4.1 Block Diagram

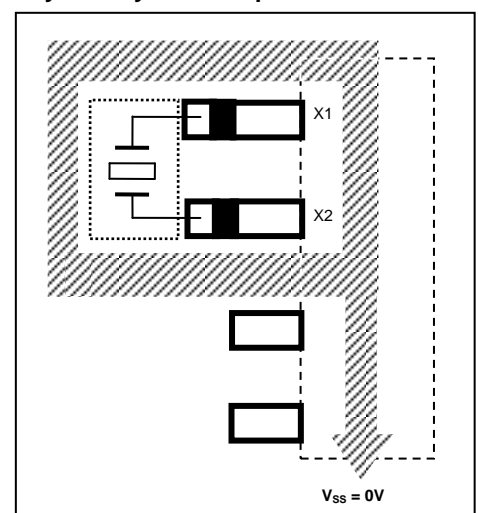


### 4.2 Application Schematic



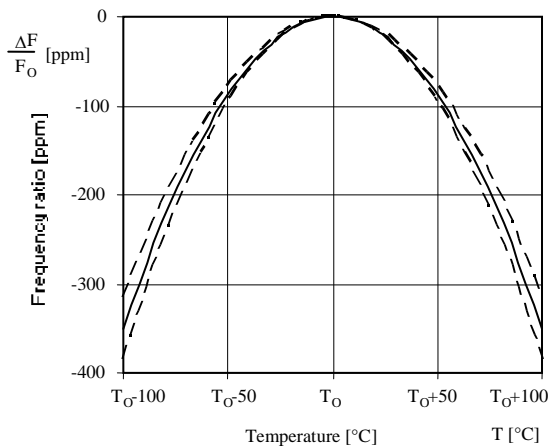
**Figure1: Application Schematic**

#### Crystal Layout Example



### 4.3 Crystal Thermal Behaviour

The frequency of the crystal is dependent on the temperature concurring with the following diagram:



**Figure 1: Crystal thermal behaviour**

$T_O$  – Turnover temperature [°C]

$F_O$  – Crystal frequency when  $T_O$  [Hz]

The following formula expresses a compensation value to be used during frequency correction.

$$\text{COMP\_val} = \text{Qcoef} \times (T - T_O)^2 - \text{XtalOffset}$$

Qcoef – Thermal quadratic coefficient [ppm/°C<sup>2</sup>]

T – Actual temperature [°C]

$T_O$  – Turnover temperature [°C]

XtalOffset – Crystal offset at  $T_O$  [ppm]

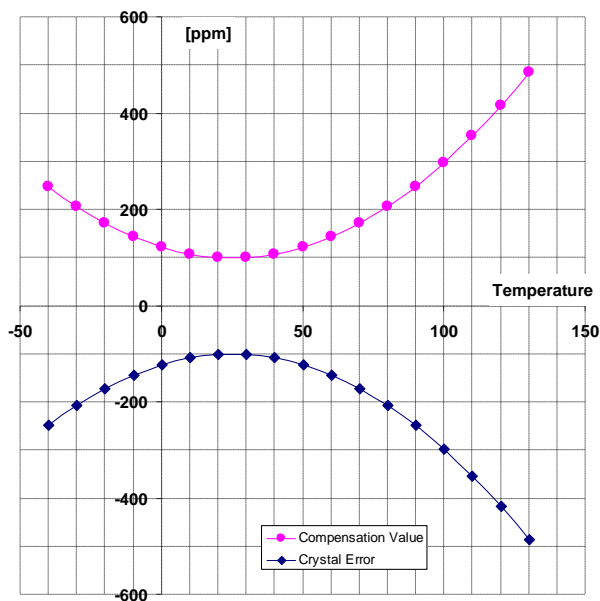
COMP\_val – Compensation value result [ppm]

The oscillator frequency is adjusted according to the equation above by using coefficients located in the EEPROM control page and the temperature.

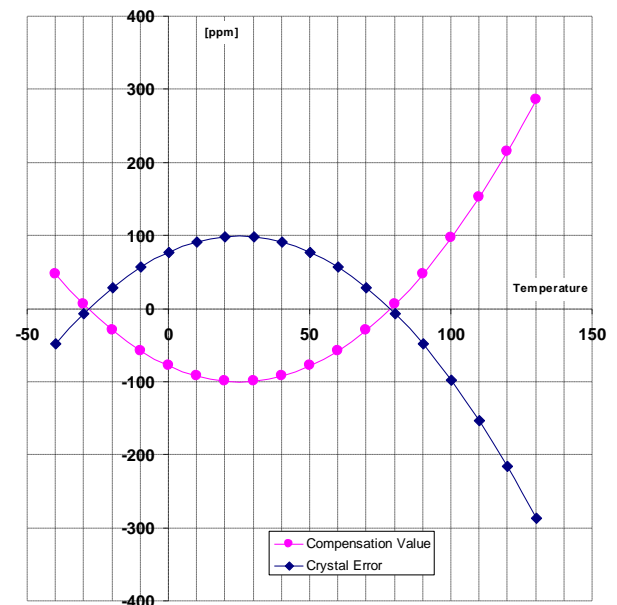
The actual temperature can be obtained from the internal thermometer or from Temp register updated externally by an application.

The principle of the frequency compensation is based on adding/removing of pulses.

**Example 1:** Qcoef=0.035;  $T_O$ =25; XtalOffset=-100



**Example 2:** Qcoef=0.035;  $T_O$ =25; XtalOffset=+100



#### 4.4 Crystal Calibration

In order to compensate temperature dependency of the used crystal, correct values of XtalOffset, Qcoef and  $T_0$  parameters shall be stored in EEPROM Control Page. User is advised to follow these steps in order to compute the parameters in a correct way:

- 1) Supply the chip from  $V_{CC}$  pin.
- 2) Set  $FD0 = FD1 = '0'$ . Set CLKOE pin to '1'. This provides the uncompensated frequency signal from the crystal oscillator directly on pin CLKOUT.
- 3) Measure output frequency  $f_O$  at different temperatures (at least five measurements in equidistant points in the whole desired temperature range are recommended). Please note that quartz crystal needs few minutes to stabilise its frequency at a given temperature.
- 4) Compute frequency deviation  $f_{err}$  in ppm of output frequency  $f_O$  from the ideal (target) frequency  $f_L = 32.768\text{Hz}$  in all measured points as follows:  
$$f_{err} = 10^6 (f_O - f_L) / f_L .$$

- 5) Find a quadratic regression of the measured dependency in form:

$$f_{err} = -c_1(T - c_2)^2 + c_3 \text{ or } f_O = aT^2 + bT + c.$$

- 6) Then real values of the searched parameters can be obtained from the following relations:

$$Qcoef_{real} = c_1 = -a,$$

$$T_{0\_real} = c_2 = -b/(2a),$$

$$XtalOffset_{real} = c_3 = c - b^2/(4a).$$

- 7) The values to be stored in EEPROM Control Page have to be corrected in the following way:

$$Qcoef = 4096 * (1.05 * Qcoef_{real}),$$

$$T_0 = T_{0\_real} - 4,$$

$$XtalOffset = 1.05 * XtalOffset_{real}.$$



## 5 Memory Mapping

Address			Description	Range	bit7	bit6	bit5	bit4	bit3	bit2	bit1	bit0
Page	Addr	Hex										
[6..3]	[2..0]											
Control Page												
00000	000	0x00	OnOffCtrl		Clk/Int	TD1	TD0	SROn	EERefOn	TROn	TiOn	WaOn
			Default		1	0	0	1	1	0	0	1
	001	0x01	IRQctrl					SRIntE	V2IntE	V1IntE	TIntE	AlntE
			Default					0	0	0	0	0
	010	0x02	IRQflags	----				SRF	V2F	V1F	TF	AF
	011	0x03	Status	----	EEBusy		PON	SR	VLOW2	VLOW1		
100	0x04	RstCtrl	----				SYSRes					
Watch Page												
00001	000	0x08	Watch Seconds	0 – 59 BCD		Seconds Tens			Seconds Units			
	001	0x09	Watch Minutes	0 – 59 BCD		Minutes Tens			Minutes Units			
	010	0x0A	Watch Hours	0 - 23 BCD 1 - 12 BCD		S12/24	pm/2	Hours Tens	Hours Units			
	011	0x0B	Watch Date	1 – 31 BCD			Date Tens		Date Units			
	100	0x0C	Watch Days	1 – 7 BCD						Days Units		
	101	0x0D	Watch Months	1 – 12 BCD				Months Tens	Months Units			
	110	0x0E	Watch Years	0 – 79 BCD		Years Tens			Years Units			
Alarm Page												
00010	000	0x10	Alarm Seconds	0 – 59 BCD	SecEq	Seconds Tens			Seconds Units			
	01	0x11	Alarm Minutes	0 – 59 BCD	MinEq	Minutes Tens			Minutes Units			
	010	0x12	Alarm Hours	0 - 23 BCD 1 - 12 BCD	HourEq		pm/2	Hours Tens	Hours Units			
	011	0x13	Alarm Date	1 – 31 BCD	DateEq		Date Tens		Date Units			
	100	0x14	Alarm Days	1 – 7 BCD	DayEq					Days Units		
	101	0x15	Alarm Months	1 – 12 BCD	MonthEq			Months Tens	Months Units			
	110	0x16	Alarm Years	0 – 79 BCD	YearEq	Years Tens			Years Units			
Timer Page												
00011	000	0x18	Timer low byte	0-255	-	-	-	-	-	-	-	-
	001	0x19	Timer high byte	0-255	-	-	-	-	-	-	-	-
Temperature Page												
00100	000	0x20	Temp	-60-195 °C	-	-	-	-	-	-	-	-
EEPROM Data Page - Configuration Registers												
00101	000	0x28	EEData	----	EEPROM user data (2 bytes)							
	001	0x29										
EEPROM Control Page - Configuration Registers												
00110	000	0x30	EEctrl	----	R80k	R20k	R5k	R1.5k	FD1	FD0	ThEn	ThPer
			Default	0	0	0	0	0	0	1	0	
	001	0x31	XtalOffset	±121	sign	-	-	-	-	-	-	-
			Default	-	-	-	-	-	-	-	-	
	010	0x32	Qcoef	----	-	-	-	-	-	-	-	-
			Default	-	-	-	-	-	-	-	-	
	011	0x33	TurnOver	4-67 °C			-	-	-	-	-	-
			Default			-	-	-	-	-	-	
RAM Page (User data RAM)												
00111	000-111	0x38-0x3F	RAMdata	----	8 bytes of data							

**Table 8**


Unused bit (Read as zero; write has no influence)

**Notes and Settings:**

- Only pages 0 to 7 are used. Unused pages are for test purposes. The application should not write into unused pages and addresses.
- The crystal offset must be set to within  $\pm 121$  ppm.
- Zero values are read from unused addresses.
- Watch, Alarm, Timer pages have to be set by an application before use.
- The bit 7 (MSB) of the Alarm registers (SecEq, MinEq.) have to be set to '1' to perform the comparison. (See paragraph 8.3)

## 6 Definitions of terms in the memory mapping

**Control Page - Register *OnOffCtrl***

Clk/Int	Selects if clock or interrupt is applied onto the $\overline{\text{IRQ}}/\text{CLKOUT}$ pin ('0' = $\overline{\text{IRQ}}$ output; '1' = CLKOUT output) – CLKOUT output is the default state after reset
TD0, TD1	Selects decrement rates for Timer (32 Hz after reset)
SROn	Enables Self-Recovery function (ON after reset)
EERefOn	Enables Configuration registers refresh each 1 hour (ON after reset)
TROn	Enables Timer Auto-reload mode ('0' – reload disabled; '1' – reload enabled)
TiOn	Enables Timer (OFF after reset)
WaOn	Enables 1 Hz clock for Watch (ON after initialisation)

**Control Page - Register *IRQctrl***

SRIntE	Self-Recovery interrupt enable
V2IntE	VLOW2 interrupt enable
V1IntE	VLOW1 interrupt enable
TIntE	Timer interrupt enable
AlntE	Alarm interrupt enable

**Control Page - Register *IRQflags***

SRF	Self-Recovery interrupt flag (bit is set to '1' when Self-Recovery reset is generated)
V2F	VLOW2 interrupt flag (bit is set to '1' when power drops below Vlow2)
V1F	VLOW1 interrupt flag (bit is set to '1' when power drops below Vlow1)
TF	Timer interrupt flag (bit is set to '1' when Timer reaches ZERO)
AF	Alarm interrupt flag (bit is set to '1' when Watch matches Alarm)

**NOTE:** *Flags can be cleared by '0' writing.*

**Control Page - Register *Status***

EEBusy	EEPROM is busy (bit is set to '1' when EEPROM write or Configuration Registers refresh is in progress) (Read Only)
PON	Power ON (bit is set to '1' at Power On; clear by '0' writing)
SR	Self-Recovery reset or System reset detected (clear by '0' writing)
VLOW2	Voltage level $V_{CC}$ or $V_{Back}$ below Vlow2 level (clear by '0' writing)
VLOW1	Voltage level $V_{CC}$ or $V_{Back}$ below Vlow1 level (clear by '0' writing)

**Control Page - Register *RstCtrl***

SYSRes	System reset register; writing '1' will initiate restart of the logic (Watch, Alarm and Timer parts excluded). After the restart, status bit SR is set. The register is cleared after restart of the logic.
--------	---

**Watch Page - Registers *Watch Seconds, Watch Minutes, Watch Hours, Watch Date, Watch Days, Watch Months, Watch Years***

Watch information (BCD format)	
S12/24	12-hours or 24-hours format selection; 12-hours: S12/24 = '1', 24-hours: S12/24 = '0'
PM/2	S12/24 = '0' PM/2 represents value '2' of tens, S12/24 = '1' PM/2 = '1' represents PM (afternoon), PM/2 = '0' represents AM (morning)

**Alarm Page - Registers *Alarm Seconds, Alarm Minutes, Alarm Hours, Alarm Date, Alarm Days, Alarm Months, Alarm Years***

Alarm information (BCD format)	
PM/2	S12/24 = '0' PM/2 represents value '2' of tens, S12/24 = '1' PM/2 = '1' represents PM (afternoon), PM/2 = '0' represents AM (morning)

**Timer Page - Registers *TimLow*, *TimHigh***

TimLow	Timer value (Low byte)
TimHigh	Timer value (High byte)

**Temperature Page - Register *Temp***

Temp	Temperature (range from -60° C to 190°C with 0°C corresponding to a content of 60d)
------	---

**EEPROM Data Page - Register *EEData***

EEData	General purpose EEPROM data bytes
--------	-----------------------------------

**EEPROM Control Page - Register *EECtrl***

R80k, R20k, R5k, R1.5k	Selects trickle charger resistors between $V_{High}$ and $V_{Back}$
FD0, FD1	Selects clock frequency at $\overline{IRQ}/CLKOUT$ pin.
ThEn	Enables thermometer ('0' = disabled; '1' = enabled)
ThPer	Selects thermometer activation period ('0' = 1 second; '1' = 16 seconds)

**EEPROM Control Page - Register *XtalOffset***

XtalOffset	Crystal frequency deviation at Turnover temperature $T_O$ in ppm. Example: value 63d is related to 60 ppm.
------------	--

$$XtalOffset = 1.05 * XtalOffset_{real}$$

where  $XtalOffset_{real}$  is real value of crystal frequency deviation at Turnover temperature of the used crystal in ppm.

**Note:** Coefficient 1.05 (exactly 1.048576) is the result of the internally used frequency compensating method.

**EEPROM Control Page - Register *Qcoef***

Qcoef	Thermal quadratic coefficient of the crystal. Example: value 151d is related to 0.035 ppm/°C²,
-------	--

$$Qcoef = 4096 \times 1.05 \times Qcoef_{real},$$

where  $Qcoef_{real}$  is real value of thermal quadratic coefficient of the crystal in ppm/°C².

**EEPROM Control Page - Register *TurnOver***

TurnOver	Turnover temperature of the crystal (values 0 to 63d are related to temperature 4 to 67 °C). Example: value 21d is related to 25°C.
----------	---

$$T_0 = T_{0\_real} - 4,$$

where  $T_{0\_real}$  is real value of Turnover temperature of the crystal in °C.

**RAM Page - Register *RAMdata***

RAMdata	General purpose RAM data bytes
---------	--------------------------------

## 7 Serial communication

Depending on the EM3027 version, the serial communication is performed in I2C or SPI mode.

A serial communication with the EM3027 starts with a "Transmission START" and terminates with the "Transmission STOP".

### "Transmission START"

I2C – START condition  
SPI – CS goes to '1'

### "Transmission STOP"

I2C – STOP condition  
SPI – CS goes to '0'

When the "Transmission START" is detected, a copy of the content of the addressed Watch-, Alarm-, Timer- and Temperature-register is stored into a cache memory. Data for a following read access are provided from this cache memory. Data in the cache memory are stable until the "Transmission STOP".

During a write access, data are written into the cache memory.

When the "Transmission STOP" of a WRITE transmission is detected, the content of modified registers in the cache memory is copied back into the Watch-, Alarm-, Timer- and/or Temperature-register.

### 7.1 How to perform data transmission through I2C

The I2C protocol is a bidirectional protocol using 2 wires for master-slave communication: SCL (clock) and SDA (data). The two bus lines are driven by open drain outputs and pulled up externally. MSB is sent first.

The communication is controlled by the master. To start a transmission, the master applies the START condition and generates the SCL clocks during the whole transmission. The master terminates the transmission by sending the STOP condition.

The first byte contains the 7 bit slave address and the R/W bit. The slave address must correspond to the fixed slave address of the EM3027. After each byte, the receiver outputs an acknowledge bit ACK to confirm correct receipt of the byte by a '0' level.

At a WRITE transmission (R/W = 0), the master sends slave address, register address and data bytes.

In the EM3027, the upper 5 bits of a register address form a "page address", the 3 lower bits are an auto-incrementing sub-address. The "page-address" is defined by a WRITE transmission. During a transmission, the 3 lower address bits are internally incremented after each data byte.

At a READ transmission (R/W = 1), the slave sends data and the master gives the ACK bit(s). The "page-address" shall be defined by a WRITE transmission, completed with the STOP condition.

The 3 lower address bits are incremented when an ACK is received.

If ACK is not received, no auto-increment of the address is executed and a following read outputs data of the same address.

The EM3027 works as slave. Its slave address is fixed to '1010110'.

#### I2C: Write transmission

Slave Address		R/W										
S	1010110	0	ACK <sub>s</sub>	Address	ACK <sub>s</sub>	Data Byte (1)	ACK <sub>s</sub>	Data Byte (n-1)	ACK <sub>s</sub>	Data Byte (n)	ACK <sub>s</sub>	P

#### I2C: Read transmission

Slave Address		R/W					Slave Address		R/W						
S	1010110	0	ACK <sub>s</sub>	Address	ACK <sub>s</sub>	P	S	1010110	1	ACK <sub>s</sub>	Data byte (1)	ACK <sub>m</sub>	Data byte (n)	ACK <sub>m</sub>	P

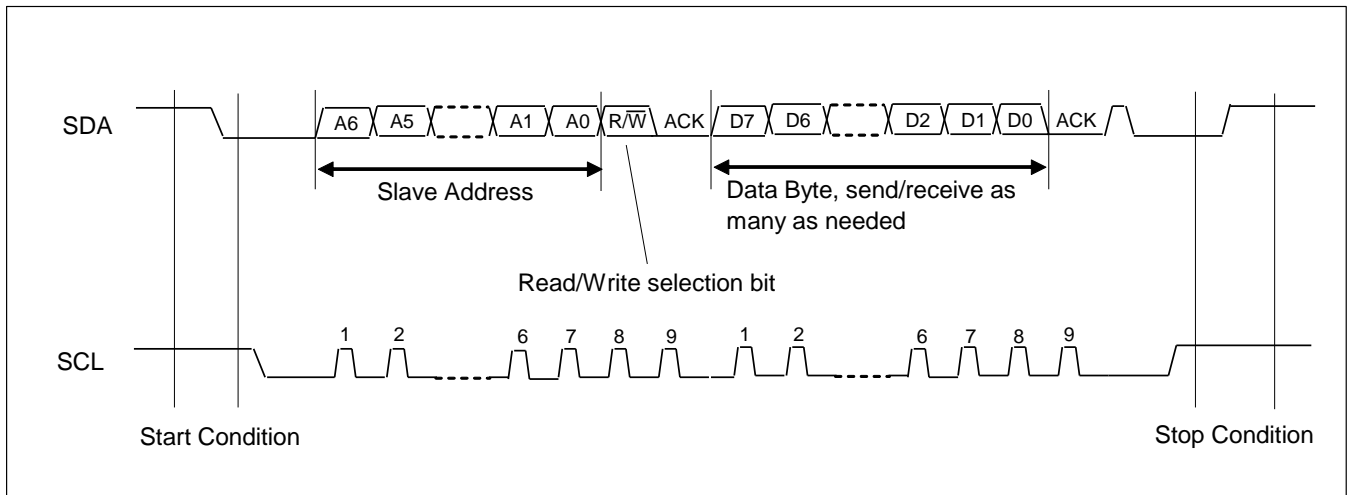
S ... start condition sent by the master

ACK<sub>s</sub> ... acknowledge from the receiver (slave)

ACK<sub>m</sub> ... acknowledge from the receiver (master)

R/W ... read/write select ('0': master writes data)

P ... stop condition



**Figure 2: I2C Communication**

Noise suppression circuitry is implemented rejecting spikes shorter than 50ns on SCL and SDA bus lines.

## 7.2 How to perform data transmission through SPI

The SPI interface connects master and slave circuits.

4 connections are used: CS = Chip Select, SCK = Serial Clock, SI = Serial Data Input and SO = Serial Data Output.

SPI is a byte oriented protocol with MSB first mode. Data are changing on SCK falling edge and sampled on rising edge.

A transmission is started by the master by rising the CS input of the selected slave to '1'. The transmission is terminated by the master by putting '0' level the CS input.

The first bit is the R/W bit, R/W = '0' means a WRITE transmission, where the master sends the data via the SI line. R/W = '1' defines a READ transmission, where the slave outputs the data on the SO line.

The following 7 bits of the first byte form the address of the register in the EM3027, where the data are written or read. (MSB is first bit at position 2 in this address byte.) The not transmitted 8<sup>th</sup> bit of the register address is set internally to '0'.

In the EM3027, the upper 5 bits of an address form a "page address", the 3 lower bits are an auto-incrementing sub-address. The "page-address" is defined by a WRITE transmission. During a transmission, the 3 lower address bits are incremented internally after each byte.

During a WRITE transmission, the master defines the register address and sends then data bytes, using the auto-increment of the lower address part (bit 2 to 0) within the EM3027.

The page address is fixed until a new transmission is started.

SO data output of EM3027 is in Hi-Z state during the WRITE transmission.

If READ transmission is initiated, data are output after the address byte by the EM3027.

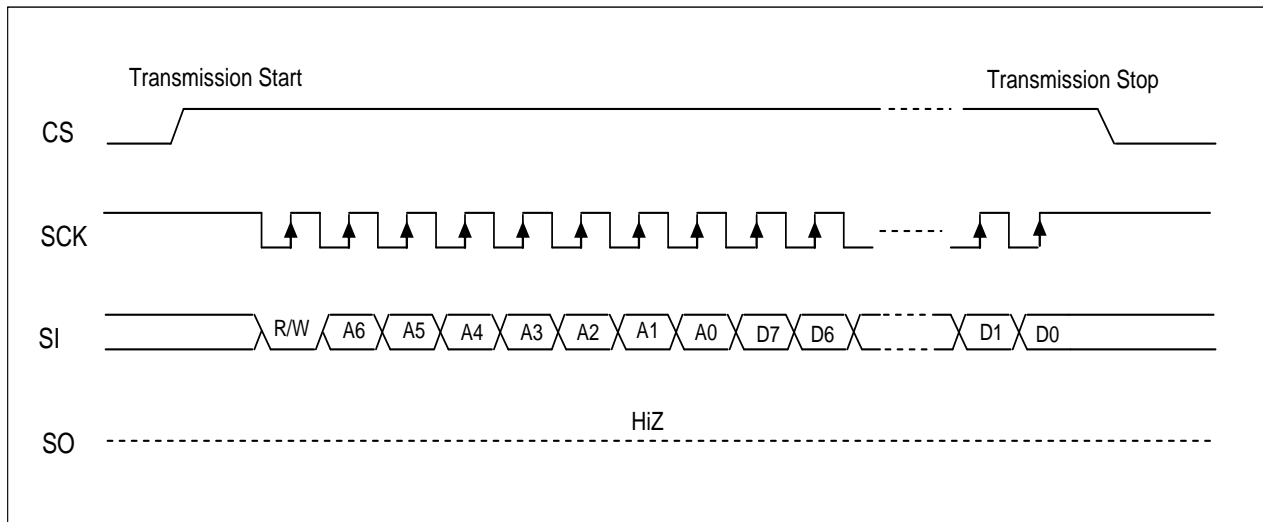
The lower part of the address (bit 2 to 0) is automatically incremented after each data byte. The page address is not changed until a new transmission is started.

SO is in Hi-Z while the address byte is sent. During data output by SO, the SI input has no influence.

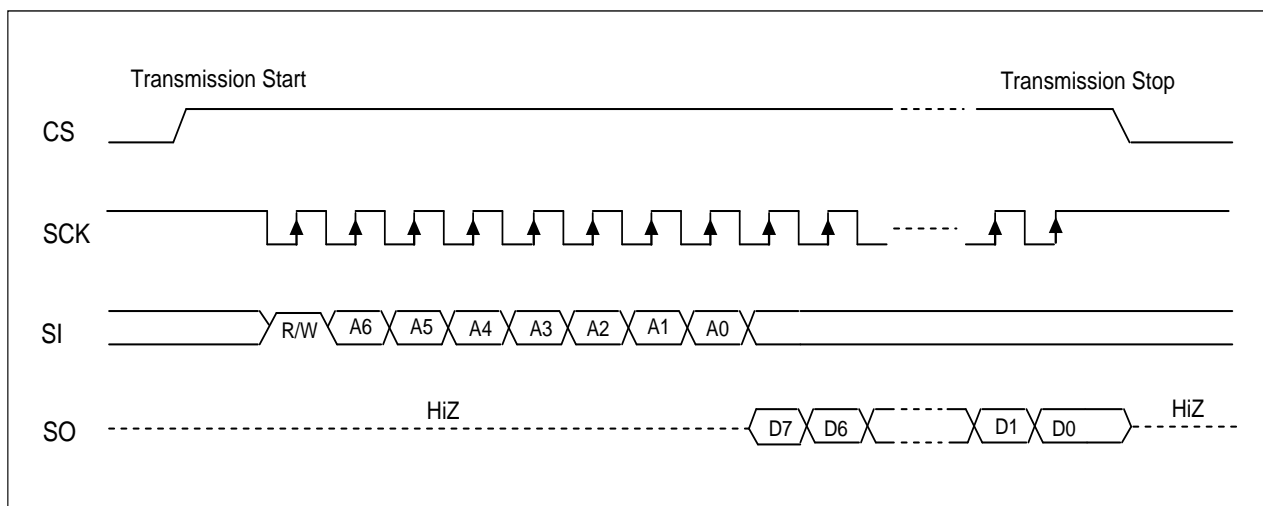
When CS is at '0' level, SO is Hi-Z and SCK, SI can be left floating.

SO and SI can be connected together to form a 3-wire interface with CS, SCK and Serial Data Input/Output.

The EM3027 works as slave. The CS input has a pull-down resistor of 100 kΩ.



**Figure 3: SPI Write Transmission**



**Figure 4: SPI Read Transmission**

## 8 Functional Description

### 8.1 Start after power-up

- A The chip is in reset state when the supply voltage is below an internal threshold level (PON in Status register 0x03 goes to '1'). When the supply level is higher than this threshold voltage, the reset is released.
- B When the supply voltage is higher than the oscillator start-up voltage, the basic clocks for Watch and control logic become active after the oscillator start time.
- C With clocks present, the voltage detector starts in fast mode to measure the supply voltage. When a voltage higher than Vlow2 is detected, the fast detection mode is stopped and the EEPROM read is enabled.
- D Configuration registers are loaded with the configuration data read from the EEPROM (Addresses from 0x28 to 0x33).
- E If thermometer is enabled (ThEn='1' and VLOW1='0'), temperature is measured and compensation value for frequency correction evaluated.
- F The EM3027 starts its normal function, depending on the supply voltage level applied.

### 8.2 Normal Mode function

The chip has following functions in Normal Mode:

1. **Voltage detection** – The voltage detection is executed each second.
2. **Temperature measurement** – It is executed, if thermometer is enabled (ThEn='1') and VLOW1='0'.
3. **Frequency compensation** – The compensation of the oscillator frequency works continuously.
4. **Configuration Registers refresh** – The EEPROM is read each hour to refresh the content of the configuration registers (supply voltage must be above Vlow2 for EEPROM read).
5. **Watch/Alarm** – The Watch function is continuously active, whereas the Alarm function depends on its activation.
6. **Timer** – Is active when enabled.
7. **Self-Recovery system** – Is enabled by default (can be disabled by the application).
8. **Serial interface** – The communication works if  $V_{CC} > V_{CC\_min}$  and  $V_{CC} > V_{Back}$ .

### 8.3 Watch and Alarm function

The Watch part provides timing information in BCD format. The timing data is composed of seconds, minutes, hours, date, weekdays, months and years. The corresponding values are updated every second.

The Watch part setup is provided by Write transmission into the Watch Page (Address 0x08h to 0x0Eh). After the transmission, the Watch is restarted from the setup values after one second.

The Alarm function is activated by setting and enabling the alarm registers (Address 0x10h to 0x16h). Each Alarm byte has its own enable bit. It is the bit 7. Recommended combinations of enabled bits are described in the table below.

SecEq	MinEq	HrsEq	DateEq	DaysEq	MonthEq	YearEq	Al_period
1	0	0	0	0	0	0	min
1	1	0	0	0	0	0	hrs
1	1	1	0	0	0	0	day
1	1	1	1	0	0	0	month
1	1	1	1	0	1	0	year
1	1	1	0	1	0	0	week

**Table 9: Alarm Period Selection**

- Both Watch and Alarm parts must be set by an application before use
- The bits SecEq to YearEq enable the comparison of the corresponding registers

#### 8.4 Timer function

The 16-bit count down timer can be enabled/disabled by TiOn bit.

The timer input frequency is selected by TD1, TD0 bits according to the following table:

TD1	TD0	Timer frequency
0	0	32 Hz
0	1	8 Hz
1	0	1 Hz
1	1	0.5 Hz

**Table 10: Timer Frequency Selection**

The timer can run in Zero-Stop or Auto-Reload mode (TROn bit: '0' = Zero-Stop mode, '1' = Auto-Reload mode).

When TROn = '0', then it is possible to read current value of the timer. If TROn = '1', then last written value is read from cache memory. The value in the cache memory is used as the new value for reloading (Auto-Reload mode).

Frequency selection (TD1, TD0) and mode selection (TROn) can be written only when the timer is stopped (TiOn = '0'). Timer values (TimLow, TimHigh) can be written only when the timer is stopped (TiOn = '0' and TROn = '0').

**NOTE:** The "Timer Page" can also be used as a general purpose register when the timer function is not used.

#### 8.5 Temperature measurement

The integrated thermometer has a resolution of 1°C.

The thermometer is disabled when ThEn = '0' and enabled when ThEn = '1'. By default, the thermometer is enabled. Thermometer period is selectable by ThPer bit according to the table below:

ThPer	Period in Seconds
0	1 s
1	16 s

**Table 11: Thermometer Period**

The thermometer is automatically disabled when VLOW1 status bit is at '1'.

When the thermometer is disabled (ThEn = '0'), the Temp register can be written. Temp register uses a cache memory to keep stable value during a whole transaction (read/write).

#### 8.6 Frequency compensation

There is a frequency compensation unit (FCU) inside EM3027. FCU compensates quartz crystal native frequency in dependency on actual compensation value (COMP\_val).

FCU is always running.

During chip power-up, if ThEn = '1' and VLOW1 = '0' temperature measurement is enabled and COMP\_val is computed. Otherwise, COMP\_val is set to 0 ppm.

In Normal mode, new compensation value is computed each 32 seconds. The only exception is when ThEn = '1' and VLOW1 = '1'. In this case, temperature measurement and COMP\_val computation are blocked and FCU uses the last computed compensation value.

For the evaluation of COMP\_val, actual content of Temp register (0x20) is used. The compensation value is computed according to the equation described in **paragraph 4.3**.

Content of Temp register is updated either after a temperature measurement (when ThEn = '1' and VLOW1 = '0') or after Temp register write transaction (when ThEn = '0'). After power-up content of Temp register is undefined.

If thermometer is disabled (ThEn = '0') user is advised to periodically update Temp register with actual ambient temperature in order to have correct input data for COMP\_val computation.



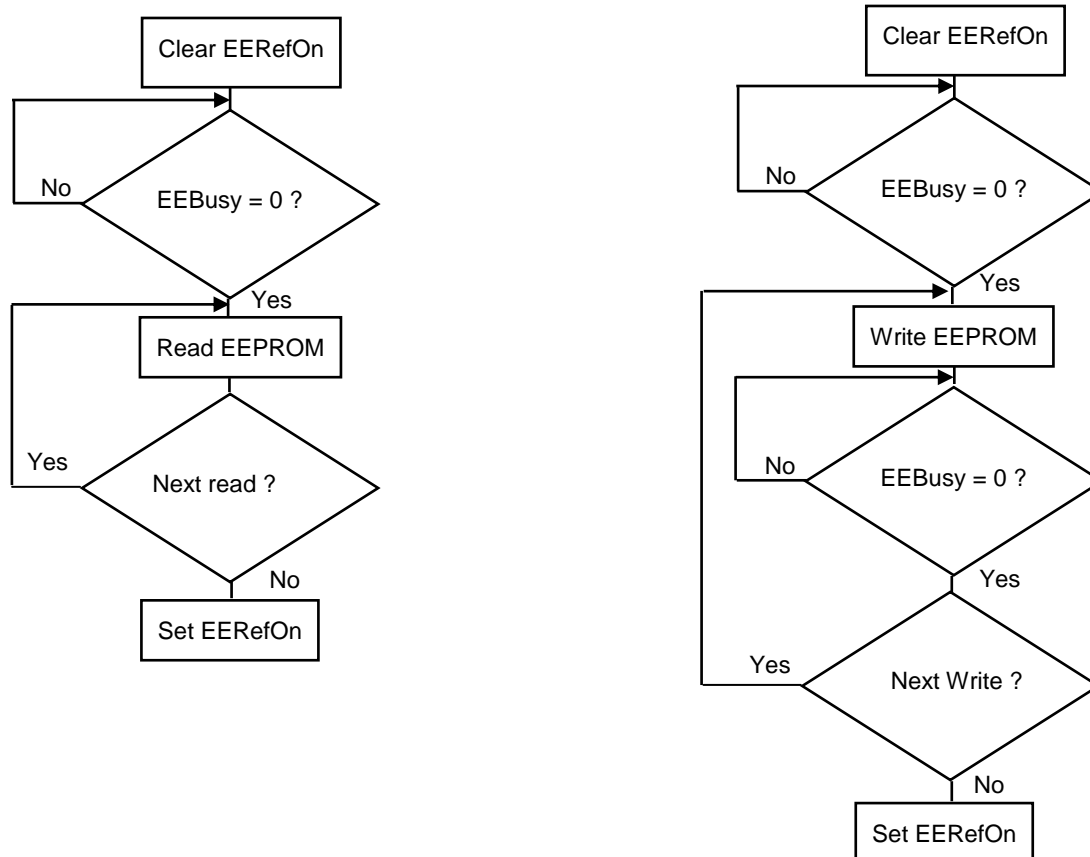
## 8.7 EEPROM memory

Before any EEPROM access (read/write), the bit EERefOn has to be cleared by the application to prevent from access collision with the Configuration Registers.

Then the application has to read EEBusy bit and if EEBusy = '0', then EEPROM access can be started.

After the write command (at "Transmission STOP") the current state of EEPROM writing is monitored by EEBusy register bit at '1'. EEBusy goes to '0' when EEPROM writing is finished.

**NOTE:**  $V_{CC}$  must be applied during the whole EEPROM write (i.e. until EEBusy = '0') and must be higher than  $V_{prog}$ .



After EEPROM write command, EEBusy bit is set to '1' for a time period depending on that how many bytes of EEPROM Data or Control page were written:

EEPROM Write Operation	EEBusy bit set to '1' for [ms]
Any write operation into EEPROM Data page (one or two bytes within one transaction)	35
Single byte write operation into EEPROM Control page	97
Multiple byte write operation into EEPROM Control page (two, three or four bytes within one transaction)	<135

### 8.7.1 EEPROM Control Page

This part is composed of 4 bytes purposed for miscellaneous function control and for crystal compensation constants.

EEctrl byte contains: trickle charger selectors (R80k, R20k, R5k, R1.5k); output clock frequency selector (FD1, FD0); thermometer enable and thermometer period selector.

### 8.7.2 Clock Output

Output clock frequency is selected by FD1, FD0 bits in EEctrl register.

FD1	FD0	Select Clock Output	Description
0	0	32.768 kHz	From crystal oscillator, without frequency compensation
0	1	1024 Hz	With frequency compensation
1	0	32 Hz	
1	1	1 Hz	

**Table 12: Output Clock frequency selection**

### 8.7.3 Configuration Registers

All the configuration data from EEPROM (i.e. EEctrl, XTalOffset, Qcoef, TurnOver, EEData) is hold in configuration registers.

Data from EEPROM is loaded to these registers during power-up sequence and is refreshed each hour, if 'Configuration Registers refresh' feature is enabled (EERefOn = '1').

Regular refresh of Configuration Registers prevents their content to be corrupted by strongly polluted electrical environment (EMC problems, disturbed power supply, etc.).

It is recommended to enable 'Configuration Registers refresh' feature.

### 8.7.4 EEPROM User Memory

Two bytes of the memory are dedicated for the application (addresses 0x28 and 0x29).

### 8.8 RAM User Memory

RAM user memory size is 8 bytes (addresses 0x38 to 0x3F). The state of the RAM data after power-up is undefined.

### 8.9 Status Register

The purpose of EEBusy bit is to inform the user about current status of the EEPROM operations.

EEBusy – status of EEPROM controller (if EEBusy = '1', then Configuration Registers refresh or EEPROM write is in progress)

The purpose of the following status bits is to record status of power supply voltage and Self-Recovery system/System reset behaviour.

- PON – status of Power-ON
- VLOW1 – status of Vlow1 voltage detection
- VLOW2 – status of Vlow2 voltage detection
- SR – status of the Self-Recovery system/System reset

If one of these status bits is set, it can be cleared only by writing '0', there is no automatic reset if the set condition disappears.

### 8.10 Interrupts

There are five interrupt sources which can output an interrupt on (INT and/or IRQ/CLKOUT) pins. The request is generated when at least one of IRQflags goes to '1' (OR function).

- AF – interrupt is provided when Watch time reaches Alarm time settings and comparison is enabled
- TF – interrupt is provided when Timer reaches ZERO
- V1F – interrupt is provided when supply voltage is below Vlow1 (when VLOW1 status bit is set)
- V2F – interrupt is provided when supply voltage is below Vlow2 (when VLOW2 status bit is set)
- SRF – interrupt is provided when Self-Recovery system invoked internal reset (when SR status bit is set)

Each interrupt source has its own interrupt enable (AIntE, TIntE, V1IntE, V2IntE, SRIntE). When the interrupt enable is '1' then the appropriate interrupt source is enabled.

Interrupt flags (IRQflags) are cleared by '0' writing into the appropriate bit. In case of V1F, V2F and SRF bits, it is necessary to clear also the corresponding status bits (Status) after interrupt bit.

### 8.11 Self-Recovery System (SRS)

The purpose of the Self-Recovery System (SRS) is to generate an internal reset in case the on-chip state machine goes into a deadlock. The function is based on an internal counter that is periodically reset by the control logic. If the counter is not reset on time, this reset will take place. It is executed after two voltage monitoring periods at the latest, i.e. 2s or 32s (ThPer bit).

A possible source of a deadlock could be disturbed electrical environment (EMC problem, disturbed power supply, etc.).

SRS sets status bit SR and resets the internal logic, except Watch, Alarm and Timer parts (i.e. time informations are not affected). Furthermore, if the SRS interrupt is enabled (SRIntE='1'), the SRF flag is set after the internal chip reset. Note, that SROn = '1' and SRIntE = '0' after the reset.

After the internal reset, the device starts with the power-up sequence (see paragraph 8.1).

SRS is automatically enabled after power-up (SROn bit). It can be disabled by writing '0' into the SROn bit in the Control Page.

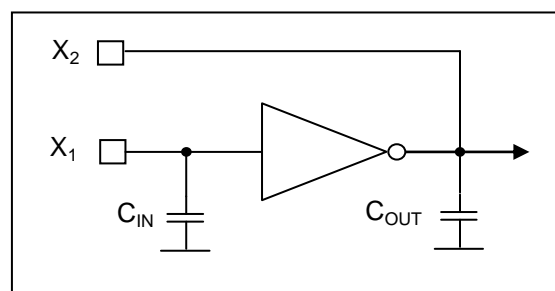
### 8.12 Register Map

The address range of the EM3027 is divided into pages. The page is addressed by the five most significant bits of the address (bits 6 ... 3). The three low significant bits of the address provide selection of registers inside the page. During address incrementing the three low significant bits (2 ... 0) are changed. The page address part is fixed during the whole data transmission.

### 8.13 Crystal Oscillator and Prescaler

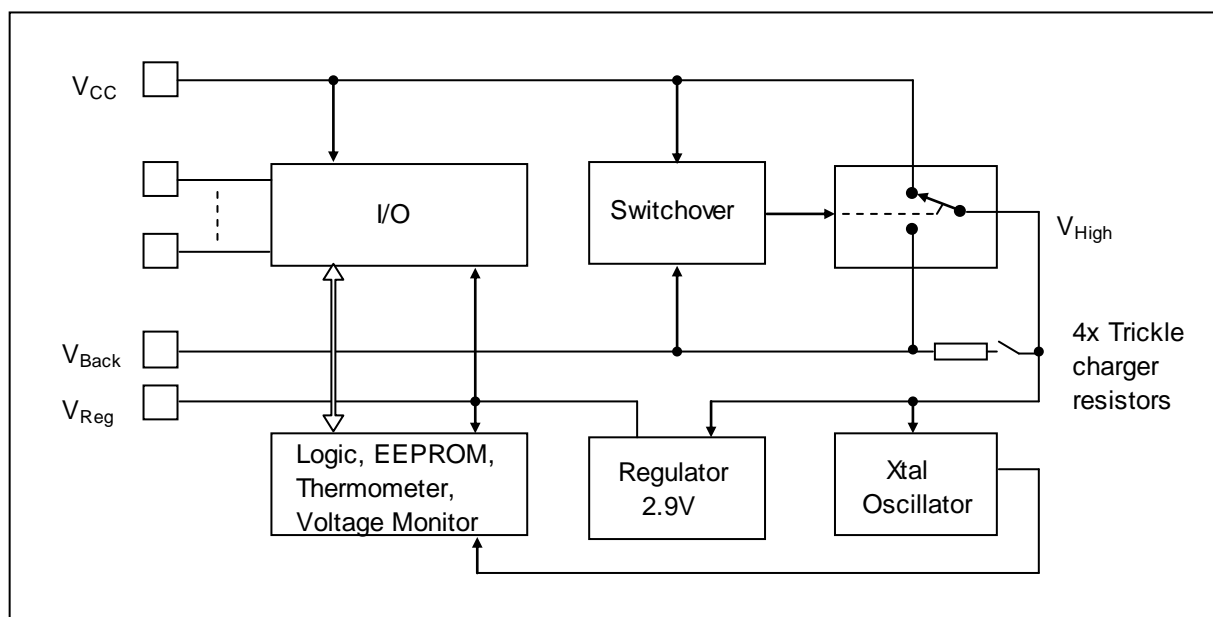
The 32.768 kHz crystal oscillator and the clock divider provide the timing signals for the functional blocks. The prescaler block is responsible for frequency division of the 32.768 kHz clock signal from the crystal oscillator. Divided frequency is then distributed between other blocks inside the chip, including Watch, Timer and control logic.

Two capacitors  $C_{IN}$  and  $C_{OUT}$  are integrated on chip – see Figure 5.



**Figure 5: Oscillator Capacitors**

## 9 Power Management



**Figure 6: Power Management**

### 9.1 Power Supplies, Switchover and Trickle Charger

The device can be supplied from the  $V_{CC}$  pin or from the  $V_{Back}$  pin.

The switchover block implemented inside the chip compares  $V_{CC}$  and  $V_{Back}$  voltages and connects the higher of them to the internal  $V_{High}$  net that supplies the chip.

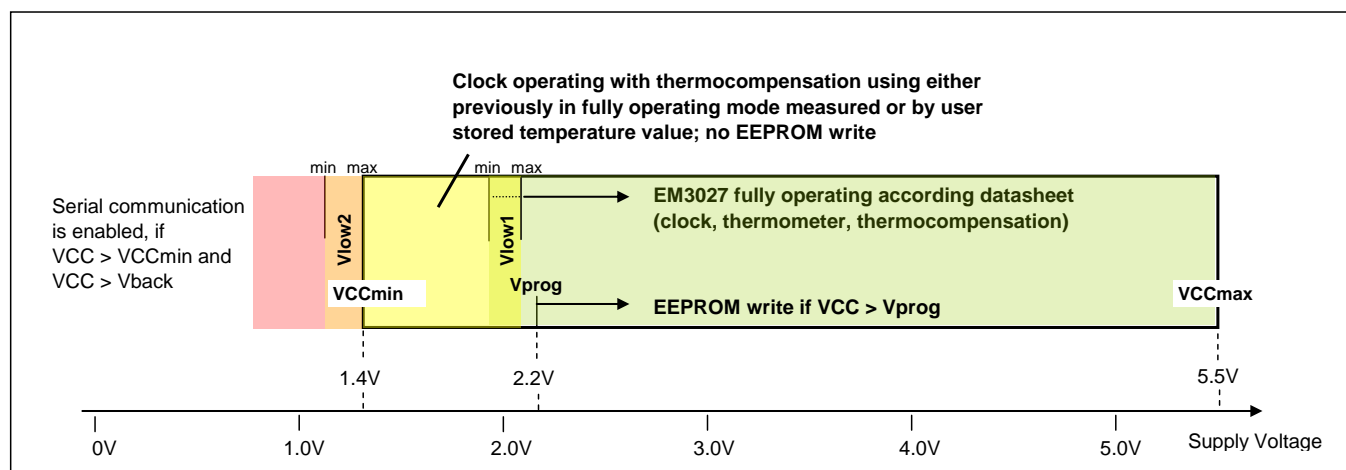
Nevertheless, the communication pins (SCL, SDA or CS, SCK, SI, SO) are supplied from the  $V_{CC}$  pin. For that reason, when serial interface (I2C or SPI) is used, the chip has to be supplied from  $V_{CC}$ . (i.e.  $V_{CC} > V_{Back}$ ).

By setting of a trickle charger bit in register  $EECtrl$ , a resistor can be inserted between  $V_{Back}$  and  $V_{High}$  voltage.

In this way, a rechargeable battery or a super-cap can be charged from the  $V_{CC}$  voltage, as long as  $V_{CC} > V_{Back}$ .

There are 4 selectable resistors connected in parallel with typical values of 80k $\Omega$ , 20k $\Omega$ , 5k $\Omega$  and 1.5k $\Omega$ . One or more resistors can be selected by  $EECtrl$  bits setting.

If a Lithium battery shall be connected to  $V_{Back}$  pin, a protection resistor of value up to 1k $\Omega$  can be connected in series with it. In this way, in case of EM3027 device damage resulting in short between both supply pins, charging current from the  $V_{CC}$  supply can be reduced to its allowed maximum level as required by UL1642 standard.



**Figure 7: EM3027 operating Voltage Areas**

## 9.2 Low Supply Detection

The supply voltage level is monitored periodically versus Vlow1 and Vlow2 levels. The monitoring rate is one second. When the voltage monitoring is running, a higher current consumption for few milliseconds occurs.

At the power-up of the device, as long as the supply voltage stays below Vlow2, the monitoring rate is accelerated. To enable normal operation, the chip must be supplied with a voltage above Vlow2, to enable the readout of initialization data from EEPROM and to stop the higher current consumption.

When the supply voltage drops from the normal range (from 2.1V to 5.5V) below Vlow1, the VLOW1 status bit is set to '1' by the voltage monitoring system.

When bit VLOW1 is at '1', the thermometer is disabled and the automatic computation of the thermal compensation value (COMP\_val) for frequency correction is inhibited. In this case, the last computed compensation value is used.

To leave the VLOW1 status, the supply voltage must be increased above the Vlow1 level and a '0' value must be written into the VLOW1 status bit via the serial interface.

When the supply voltage drops below the Vlow2 level, the VLOW2 status bit is set by the voltage monitoring system.

The VLOW2 status bit disables the read out of the EEPROM.

To leave the VLOW2 status, the supply voltage must be increased above the Vlow2 level and a '0' value must be written into the VLOW2 status bit via the serial interface.

Below Vlow2 level, device functionality is not guaranteed and register contents can be corrupted. Therefore, if VLOW2 status bit is set, it is recommended to perform system reset by writing of '1' into SYSRes bit in RstCtrl page and afterwards update content of Watch, Alarm and Timer registers.

## 10 AC Characteristics

### 10.1 AC characteristics – I2C

$V_{SS} = 0V$  and  $T_A = -40$  to  $+125^{\circ}C$ , unless otherwise specified

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
SCL Clock Frequency	$f_{SCL}$	$V_{CC} \geq 3.0V$			400	kHz
		$V_{CC} > 1.8V$			300	
		$V_{CC} > 1.4V$			100	
Bus Free Time Between STOP and START Condition	$t_{BUF}$	$V_{CC} \geq 3.0V$	0.6			$\mu s$
		$V_{CC} > 1.8V$	0.8			
		$V_{CC} > 1.4V$	1.0			
Hold Time (Repeated) START Condition	$t_{HD:STA}$	$V_{CC} \geq 3.0V$	0.3			$\mu s$
		$V_{CC} > 1.8V$	0.4			
		$V_{CC} > 1.4V$	0.5			
LOW Period of SCL Clock	$t_{LOW}$	$V_{CC} \geq 3.0V$	1.3			$\mu s$
		$V_{CC} > 1.8V$	1.7			
		$V_{CC} > 1.4V$	4.5			
HIGH Period of SCL Clock	$t_{HIGH}$	$V_{CC} \geq 3.0V$	0.6			$\mu s$
		$V_{CC} > 1.8V$	0.7			
		$V_{CC} > 1.4V$	0.9			
Setup Time START Condition	$t_{SU:STA}$	$V_{CC} \geq 3.0V$	0.3			ns
		$V_{CC} > 1.8V$	0.4			
		$V_{CC} > 1.4V$	0.5			
Data Hold Time	$t_{HD:DAT}$	$V_{CC} \geq 3.0V$	0.3			ns
		$V_{CC} > 1.8V$	0.4			
		$V_{CC} > 1.4V$	0.5			
Data Setup Time	$t_{SU:DAT}$	$V_{CC} \geq 3.0V$	0.3			ns
		$V_{CC} > 1.8V$	0.4			
		$V_{CC} > 1.4V$	0.5			
Data Valid Time	$t_{VD:DAT}$	$V_{CC} \geq 3.0V$	1.2			$\mu s$
		$V_{CC} > 1.8V$	1.5			
		$V_{CC} > 1.4V$	4.0			
Data Valid Acknowledge Time	$t_{VD:ACK}$	$V_{CC} \geq 3.0V$	0.9			$\mu s$
		$V_{CC} > 1.8V$	1.1			
		$V_{CC} > 1.4V$	3.5			
Rise Time of Both SDA and SCL Signals	$t_R$	$V_{CC} \geq 3.0V$			200	ns
		$V_{CC} > 1.8V$			300	
		$V_{CC} > 1.4V$			1000	
Fall Time of Both SDA and SCL Signals (See note 1)	$t_F$	$V_{CC} \geq 3.0V$			200	ns
		$V_{CC} > 1.8V$			300	
		$V_{CC} > 1.4V$			400	
Setup Time (Repeated) STOP Condition	$t_{SU:STO}$	$V_{CC} \geq 3.0V$	0.3			ns
		$V_{CC} > 1.8V$	0.4			
		$V_{CC} > 1.4V$	0.5			
Length of spikes suppressed by the input filter on SCL and SDA	$t_{SP}$				50	ns
Capacitive Load For Each Bus Line	$C_B$				200	pF
I/O Capacitance (SDA, SCL)	$C_{I/O}$				10	pF

**Table 13: I2C AC characteristics**

Parameters are guaranteed by design. They are not tested in production.

#### Calculation of external pull-up resistor

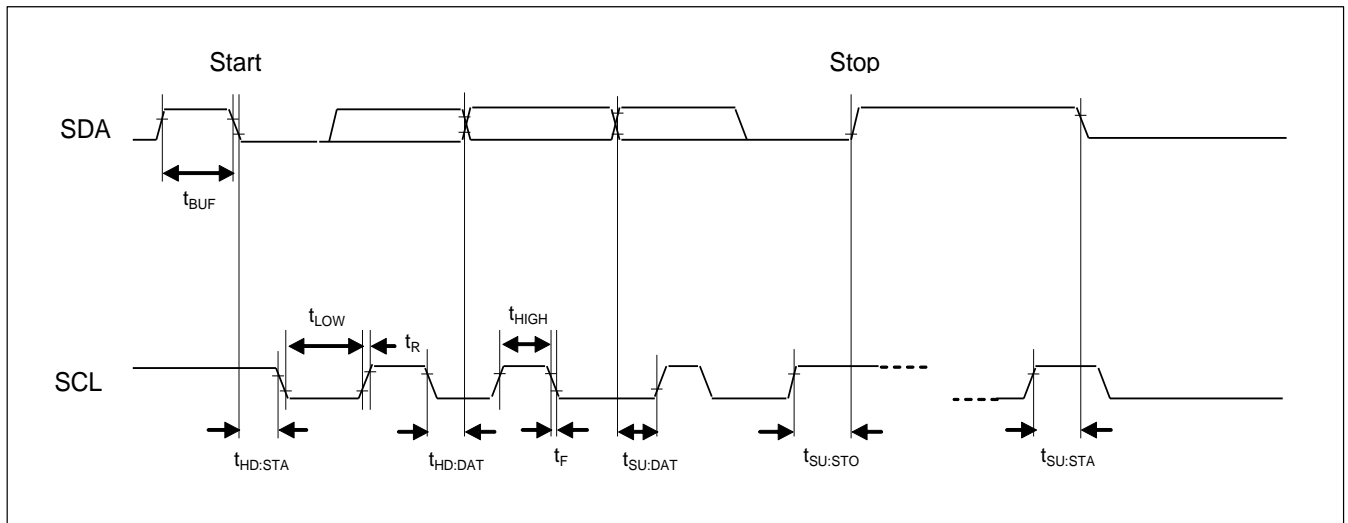
The following conditions have to be met:

Rise time is equal to  $0.847 R_{PU} (C_B + N * C_{I/O}) \Rightarrow R_{PU} < t_{R \max} / (0.847 (C_B + N C_{I/O}))$ , where N is total number of I/O pins connected to the corresponding bus line.  
( $t_R$  in ns, C in pF, R in k $\Omega$ )

The minimum value of the pullup resistor value can be calculated with the  $I_{OL}$  value of the SDA output:

$$R_{PU} = (V_{CC} - V_{OL}) / I_{OL}$$

( $I_{OL}$ : see Table 7, page 5, Output Parameters; e.g. 5mA at  $V_{CC} = 5.0V$ , with  $V_{OL} = 0.8V$  )



**Figure 8: I2C Timing**

## 10.2 I2C Specification compliance

EM3027 device with I2C serial interface was designed in compliance with Philips Semiconductors I<sup>2</sup>C-bus specification UM10204 (Rev. 03 – 19 June 2007), Fast-mode class (up to 400kbit/s). Device address consists of 7 bits. Clock stretching is not supported.

Brief manual to I2C interface read and write transmissions is to be found in §7.1.

There are, however, the following discrepancies between I2C specification and EM3027 interface:

- 1) Falling time on SDA driven by EM3027 can be shorter than  $20 + 0.1 * C_B$  ns. ( $C_B$  is total capacitive load for SDA bus line in pF) In other words, slope control of falling edges on SDA is missing.
- 2) Some timing parameters differ from the original I2C specification – refer to Table 13.

### 10.3 AC characteristics – SPI

$V_{SS} = 0V$  and  $T_A = -40$  to  $+125^{\circ}C$ , unless otherwise specified

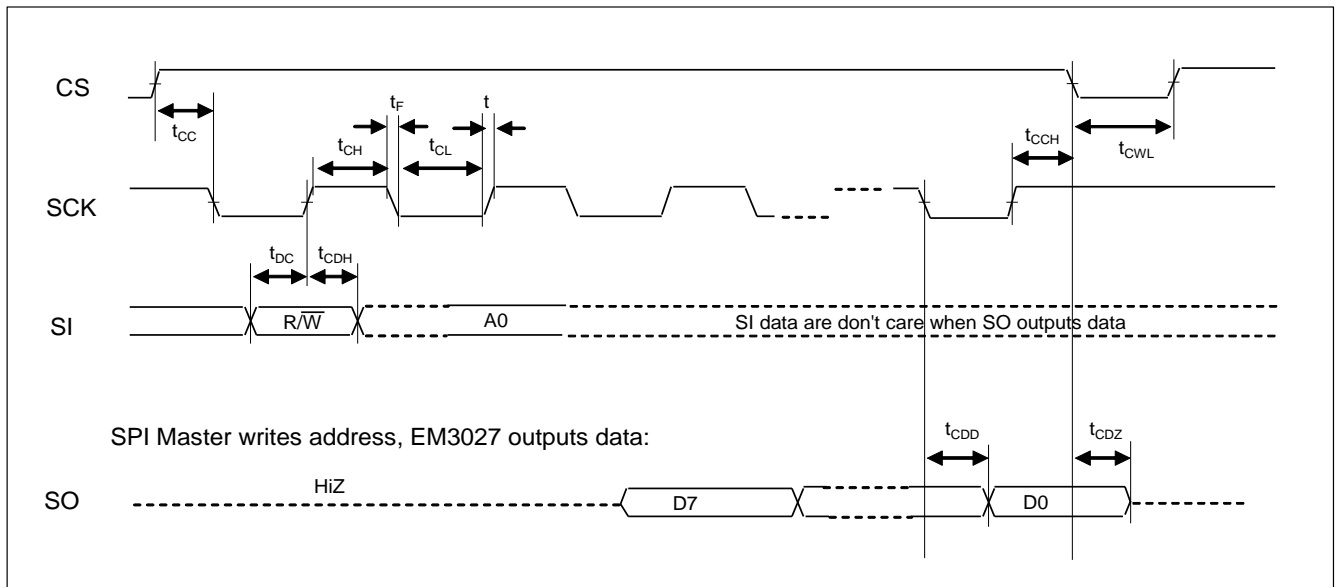
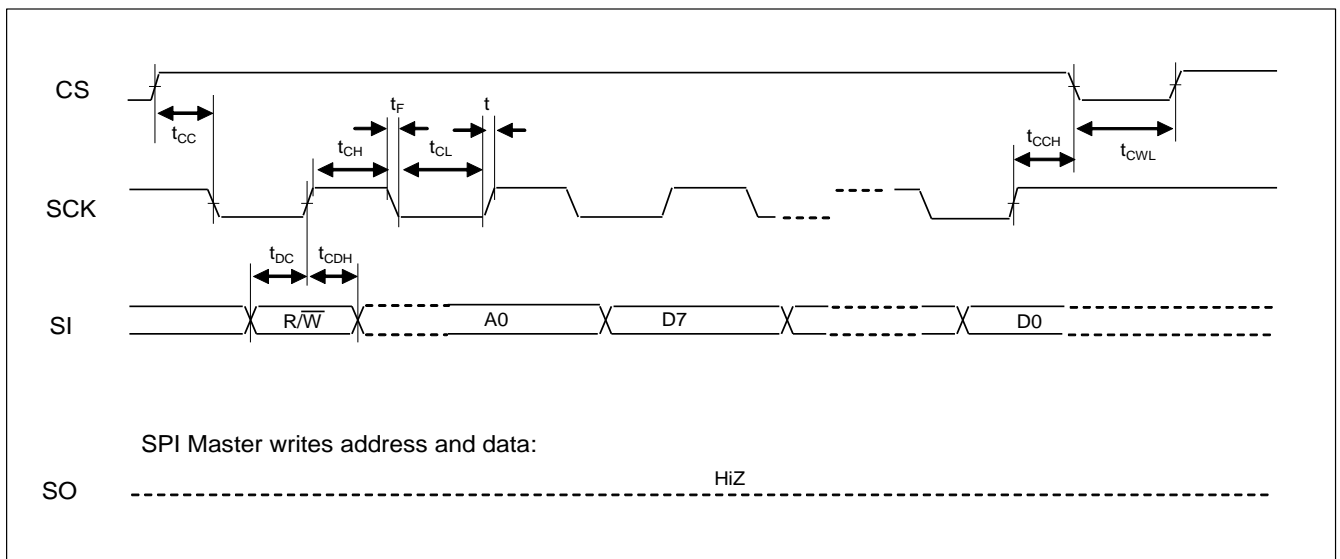
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
SCK Clock Frequency	f <sub>SCK</sub>	V <sub>CC</sub> ≥ 3.0V			1	MHz
		V <sub>CC</sub> >1.8V			600	kHz
		V <sub>CC</sub> >1.4V			200	
Data to SCK setup	t <sub>DC</sub>	V <sub>CC</sub> ≥ 3.0V	20			ns
		V <sub>CC</sub> >1.8V				
		V <sub>CC</sub> >1.4V				
SCK to Data Hold	t <sub>CDH</sub>	V <sub>CC</sub> ≥ 3.0V	200			ns
		V <sub>CC</sub> >1.8V	300			
		V <sub>CC</sub> >1.4V	500			
SCK to Data Valid	t <sub>CDD</sub>	V <sub>CC</sub> ≥ 3.0V			350	ns
		V <sub>CC</sub> >1.8V			650	
		V <sub>CC</sub> >1.4V			1300	
SCK Low Time	t <sub>CL</sub>	V <sub>CC</sub> ≥ 3.0V	400			ns
		V <sub>CC</sub> >1.8V	700			
		V <sub>CC</sub> >1.4V	1500			
SCK High Time	t <sub>CH</sub>	V <sub>CC</sub> ≥ 3.0V	400			ns
		V <sub>CC</sub> >1.8V	700			
		V <sub>CC</sub> >1.4V	1500			
SCK Rise and Fall	t <sub>R</sub> , t <sub>F</sub>	V <sub>CC</sub> ≥ 3.0V			200	ns
		V <sub>CC</sub> >1.8V			800	
		V <sub>CC</sub> >1.4V				
CS to SCK Setup	t <sub>CC</sub>	V <sub>CC</sub> ≥ 3.0V	100			ns
		V <sub>CC</sub> >1.8V				
		V <sub>CC</sub> >1.4V				
SCK to CS Hold	t <sub>CCH</sub>	V <sub>CC</sub> ≥ 3.0V	200			ns
		V <sub>CC</sub> >1.8V	300			
		V <sub>CC</sub> >1.4V	500			
CS Inactive Time	t <sub>CWL</sub>	V <sub>CC</sub> ≥ 3.0V	200			ns
		V <sub>CC</sub> >1.8V	300			
		V <sub>CC</sub> >1.4V	400			
CS to Output High Impedance	t <sub>CDZ</sub>	V <sub>CC</sub> ≥ 3.0V			50	ns
		V <sub>CC</sub> >1.8V			100	
		V <sub>CC</sub> >1.4V			200	

**Table 14: SPI AC characteristics**

Parameters are guaranteed by design. They are not tested in production.

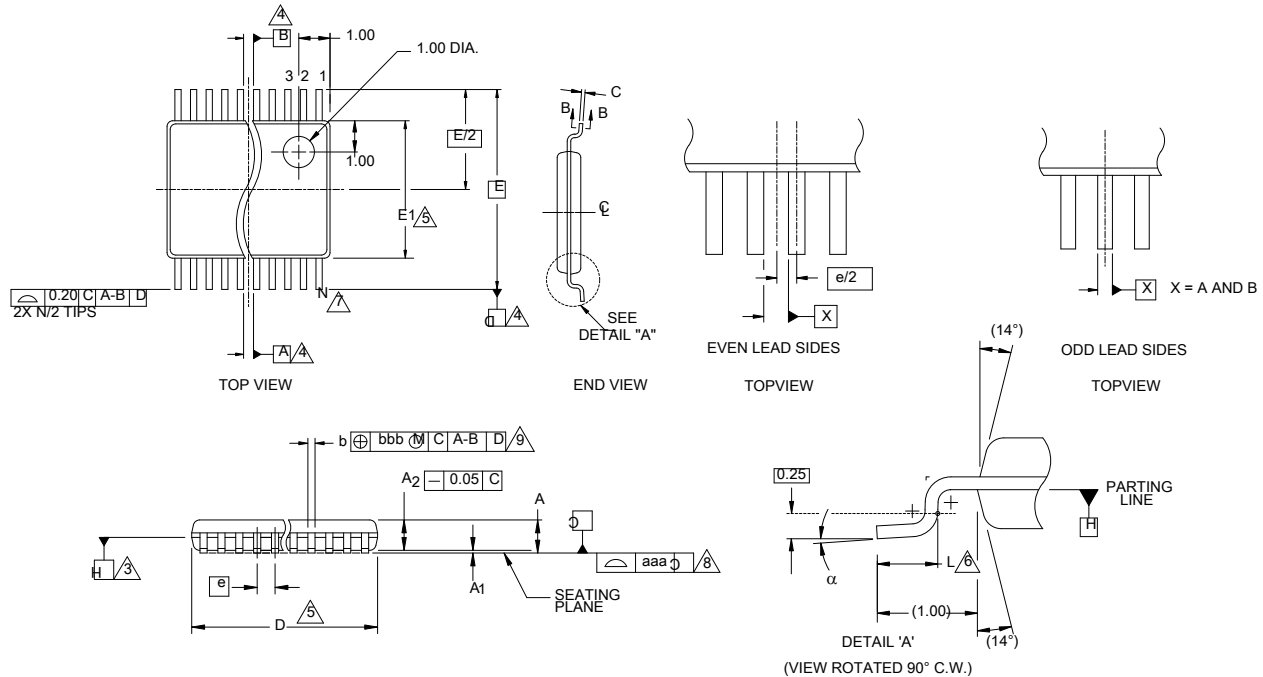
- 1) Max. bus capacitance on SO line shall be lower than 100pF when  $V_{CC} > 1.8V$  and lower than 50pF when  $V_{CC} < 1.8V$ .
- 2) Spikes on SCK signal shorter than 20ns are suppressed.




**Figure 9: SPI Read Timing**

**Figure 10: SPI Write Timing**

## 11 Package Information

### 11.1 TSSOP-08/14



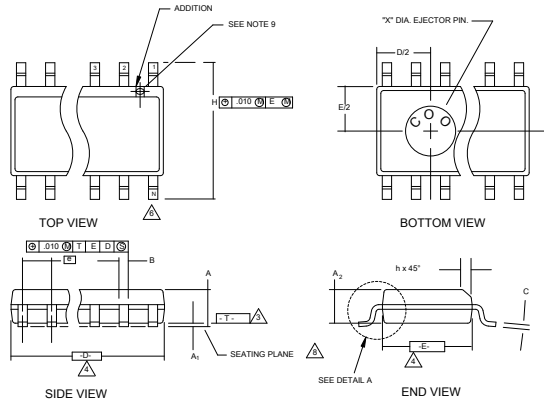
SYMBOL	COMMON DIMENSIONS			NOTE VARIATIONS	5			7	
	MIN.	NOM.	MAX.		MIN.	NOM.	MAX.	P	P1
A			1.10		2.90	3.00	3.10	1.59	3.2
A1	0.05		0.15		4.90	5.00	5.10	3.1	3.0
A2	0.85	0.90	0.95						
aaa		0.076							
b	0.19	-	0.30	9					
b1	0.19	0.22	0.25						
bbb		0.10							
c	0.09	-	0.20						
c1	0.09	0.127	0.16						
D	SEE VARIATIONS			5					
E1	4.30	4.40	4.50	5					
e		0.65 BSC							
E		6.40 BSC							
L	0.50	0.60	0.70	6					
N	SEE VARIATIONS			7					
P	SEE VARIATIONS								
P1	SEE VARIATIONS								
α	0°		8°						

ALL DIMENSIONS IN MILLIMETERS

**NOTES:**

- DIE THICKNESS ALLOWABLE IS 0.279±0.0127
  - DIMENSIONING & TOLERANCES PER ASME Y14.5M-1994.
- DATUM PLANE H LOCATED AT MOLD PARTING LINE AND COINCIDENT WITH LEAD, WHERE LEAD EXITS PLASTIC BODY AT BOTTOM OF PARTING LINE.
  - DATUM A-B AND D TO BE DETERMINED WHERE CENTERLINE BETWEEN LEADS EXITS PLASTIC BODY AT DATUM PLANE H.
  - "D" & "E1" ARE REFERENCE DATUM AND DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS, AND ARE MEASURED AT THE BOTTOM PARTING LINE. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.15mm ON D AND 0.25mm ON E PER SIDE.
  - DIMENSION IS THE LENGTH OF TERMINAL FOR SOLDERING TO A SUBSTRATE.
  - TERMINAL POSITIONS ARE SHOWN FOR REFERENCE ONLY.
  - FORMED LEADS SHALL BE PLANAR WITH RESPECT TO ONE ANOTHER WITHIN 0.076mm AT SEATING PLANE.
  - THE LEAD WIDTH DIMENSION DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.07mm TOTAL IN EXCESS OF THE LEAD WIDTH DIMENSION AT MAXIMUM MATERIAL CONDITION. DAMBAR CANNOT BE LOCATED ON THE LOWER RADIUS OR THE FOOT. MINIMUM SPACE BETWEEN PROTRUSIONS AND AN ADJACENT LEAD SHOULD BE 0.07mm

## 11.2 SO-8



	COMMON DIMENSIONS			N <sub>O</sub> E
	MIN.	NOM.	MAX.	
A	1.55	1.63	1.73	
A	0.127	0.15	0.25	
A	1.40	1.47	1.55	
B	0.35	0.41	0.49	
C	0.19	0.20	0.25	
D	4.80	4.93	4.98	3
E	3.81	3.94	3.99	
e	1.27 BSC			
H	5.84	5.99	6.20	
h	0.25	0.33	0.41	
L	0.41	0.64	0.89	
N	8			5
alpha	0°	5°	8°	
X	2.16	2.36	2.54	

THIS TABLE IN MILLIMETERS

### NOTES:

1. MAXIMUM DIE THICKNESS ALLOWABLE IS .015.
2. DIMENSIONING & TOLERANCES PER ANSI.Y14.5M - 1982.
3. "T" IS A REFERENCE DATUM.
4. "D" & "E" ARE REFERENCE DATUMS AND DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS, BUT DOES INCLUDE MOLD MISMATCH AND ARE MEASURED AT THE MOLD PARTING LINE. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.006 INCHES AT END AND .010 INCHES AT WINDOW
5. "L" IS THE LENGTH OF TERMINAL FOR SOLDERING TO A SUBSTRATE.
6. "N" IS THE NUMBER OF TERMINAL POSITIONS.
7. TERMINAL POSITIONS ARE SHOWN FOR REFERENCE ONLY
8. FORMED LEADS SHALL BE PLANAR WITH RESPECT TO ONE ANOTHER WITHIN .003 INCHES AT SEATING PLANE.
9. THE APPEARANCE OF PIN #1 I.D ON THE 8 LD IS OPTIONAL. ROUND TYPE ON SINGLE LEADFRAME AND RECTANGULAR TYPE ON MATRIX LEADFRAME.

## 12 Ordering Information

EM3027 A I D X SO8B					
<b>Part Number</b>			<b>Package</b>		
EM3027 = RTC					
<b>Switchover &amp; Trickle Charger</b>					
Not available = A			<b>SO8B=</b> 8 pin SO8 tape		
Available = B			<b>TP14=</b> 14 pin TSSOP14 tape		
<b>Interface</b>			<b>WS8=</b> Wafer sawn 8 MILS		
I2C bus = I					
SPI bus = S			<b>Functional Temperature</b>		
<b>Temperature compensation</b>			<b>Extended temperature=</b> X		
Default Temp. Compensation = D					
(Factory Standard)					

### Standard Versions

Part Number	Package	Functional Temperature	Interface	Delivery Form	Marking
EM3027AIDXWS8		-40 +125°C	I2C	Sawn 8 mils on Blue foil	
EM3027BIDXSO8B+	SO8	-40 +125°C	I2C	Tape & Reel, 2500 pcs	3027X5
EM3027BIDSSO8B+	SO8	-40 +85°C	I2C	Tape & Reel, 2500 pcs	3027 5
EM3027BSDXTP14B+	TSSOP14	-40 +125°C	SPI	Tape & Reel, 3500 pcs	3027X6
EM3027BSDSTP14B+	TSSOP14	-40 +85°C	SPI	Tape & Reel, 3500 pcs	3027 6

Please contact Sales office for other versions not shown here and for availability of non standard versions.

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